CENG4480 2021F Homework 2

Due: Nov. 07, 2021

Solutions

All solutions should be submitted to the blackboard in the format of **PDF/MS Word**.

Q1 (10%) A 8 bits ADC with full scale output voltage of 2.56V is designed to have a ±0.5 LSB accuracy. If the ADC is calibrated at 50°C and the operating temperature ranges from 0°C to 100°C, then what's the maximum net temperature coefficient of the ADC should not exceed?

A1

$$N = 8 \text{ bits}$$
 (1)

$$FSO = 2.56 \text{ V}$$
 (2)

$$1 \text{ LSB} = \frac{\text{FSO}}{2^{N-1}} \approx 10 \text{ mV} \tag{3}$$

$$\pm 0.5 \, \text{LSB} = \pm 5 \, \text{mV} \tag{4}$$

$$\alpha \times \Delta T \le \pm 5 \text{ mV} \tag{5}$$

$$\alpha \le 100 \mu V / {\rm ^{\circ}C} \tag{6}$$

Q2 (20%) Given a low-pass filter as shown in Figure 1.

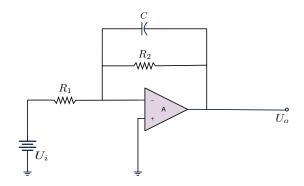


Figure 1: Q2 Circuit

- 1. If $R_1 = 10K\Omega$, $R_2 = 100K\Omega$, determine low-frequency gain $A_u(dB)$;
- 2. If cutoff frequency $f_c = 5Hz$, determine C value.

A2

$$u_o(j\omega) = -\frac{R_2//\frac{1}{j\omega C}}{R_1} u_i(j\omega) \tag{7}$$

So

$$A_u(j\omega) = \frac{u_o(j\omega)}{u_i(j\omega)} = -\frac{R_2}{R_1} \frac{1}{1 + j\omega R_2 C} = \frac{A_{u_o}}{1 + j\frac{\omega}{\omega_o}}$$
(8)

where $A_{u_o}=-\frac{R_2}{R_1}$ is low-frequency gain, $\omega_c=\frac{1}{R_2C}$ is cutoff angular frequency. If $R_1=10K\Omega,\,R_2=100K\Omega,$ low-frequency gain

$$20\log\frac{R_2}{R_1} = 20\log\frac{100}{10} = 20dB\tag{9}$$

cutoff frequency:

$$f_c = \frac{\omega_c}{2\pi} = \frac{1}{2\pi R_2 C} = \frac{1}{2\pi \times 100 \times 10^3 \times C} = 5Hz \tag{10}$$

SO

$$C = \frac{1}{2\pi \times 100 \times 10^3 \times f_c} = \frac{1}{2\pi \times 100 \times 10^3 \times 5} = 3.18 \times 10^{-7} F = 0.318 \mu F \quad (11)$$

Q3 (20%) Given the the circuit shown in Figure 2, $R_1=R_2=1k\Omega$, $R_3=10k\Omega$, $R_4=R_5=R_6=20k\Omega$, $C_1=0.05\mu F$, $C_2=0.02\mu F$, please answer the following questions.

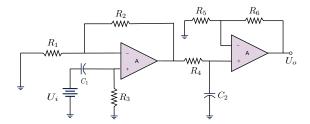


Figure 2: Q3 Circuit

- (1) Which kind of filter the circuit is?
- (2) Calculate the cutoff frequencies of the circuit. If it has both the lower and upper cutoff frequency, then you should calculate both.
- A3 (1) bandpass filter (2) The lower cutoff frequency f_{OL} is given by

$$f_{OL} = \frac{1}{2\pi R_3 C_1} = 318.3Hz \tag{12}$$

The upper cutoff frequency f_{OH} is given by

$$f_{OL} = \frac{1}{2\pi R_4 C_2} = 397.9Hz \tag{13}$$

- **Q4** A 5-bit D/A converter has a voltage output. For a binary input of 10100, an output voltage of 18 mV is produced. Determine the output voltage when the binary input is 11100.
- **A4** The binary input 10100 is equivalent to 20_{10} , then

The proportionality factor
$$=\frac{18}{20} = 0.9$$
 (14)

It follows that the output voltage is 0.9 times the binary value. The binary input 11100 is equivalent to 28_{10} , hence

Output voltage =
$$0.9 \times 28 = 25.2 mV$$
 (15)

Q5 For the DAC circuit shown in Figure 3 (using an ideal op-amp), what value of R_F will give an output range of $-10 \le V_0 \le 0V$? Assume that logic 0 = 0V and logic 1 = 5V.

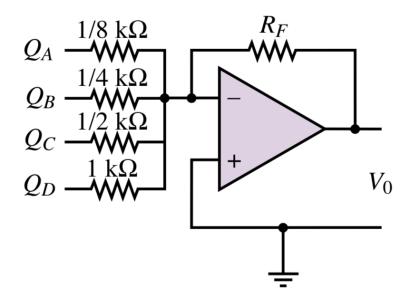


Figure 3: Q5 Circuit

A5 We have the above equation:

$$\frac{-V_0}{R_F} = \frac{(8Q_A + 4Q_B + 2Q_C + Q_D) \times 5V}{1k\Omega}$$
 (16)

when input equals 0000, $V_0 = 0V$, when input equals 1111, $V_0 = -10V$. So we get:

$$R_F = \frac{10}{15 \times 5} = \frac{2}{15} K\Omega \tag{17}$$

Q6 For the 4-bit R-2R DAC shown in Figure 4, calculate V_0 in terms of $V_{b,0} - V_{b,4}$ if V_{ref} is grounded.

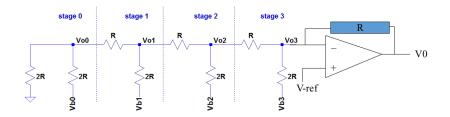


Figure 4: Q6 Circuit

A6 As shown in Figure 5, first we calculate the equivalence seen from V_{o3} ,

$$R_{eq} = R \tag{18}$$

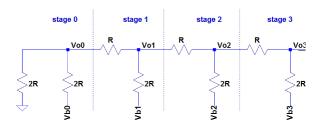


Figure 5: Load of R-2R ADC

Get contribution at V_{o3i} of each digital input V_{bi} , i=0,1,2,3 separately, it's easy to derive from Thevenin equivalent analysis,

$$V_{o30} = \frac{V_{b0}}{16} \tag{19}$$

$$V_{o31} = \frac{V_{b1}}{8}$$

$$V_{o32} = \frac{V_{b2}}{4}$$

$$V_{o32} = \frac{V_{b2}}{4}$$

$$V_{o33} = \frac{V_{b2}}{4}$$

$$V_{o33} = \frac{V_{b3}}{4}$$

$$V_{o33} = \frac{V_{b3}}{4}$$

$$V_{o33} = \frac{V_{b3}}{4}$$

$$V_{o33} = \frac{V_{b3}}{4}$$

$$V_{o34} = \frac{V_{b3}}{4}$$

$$V_{o35} = \frac{V_{b4}}{4}$$

$$V_{o35} = \frac{V_{b5}}{4}$$

$$V_{o36} = \frac{V_{b6}}{4}$$

$$V_{o37} = \frac{V_{b7}}{4}$$

$$V_{o32} = \frac{V_{b2}}{4} \tag{21}$$

$$V_{o33} = \frac{V_{b3}}{2} \tag{22}$$

(23)

then, we have,

$$V_{o3} = \frac{V_{b0}}{16} + \frac{V_{b1}}{8} + \frac{V_{b2}}{4} + \frac{V_{b3}}{2}$$
 (24)

Using the quality of op amp,

$$V_o = \frac{V_{b0}}{16} + \frac{V_{b1}}{8} + \frac{V_{b2}}{4} + \frac{V_{b3}}{2}$$
 (25)