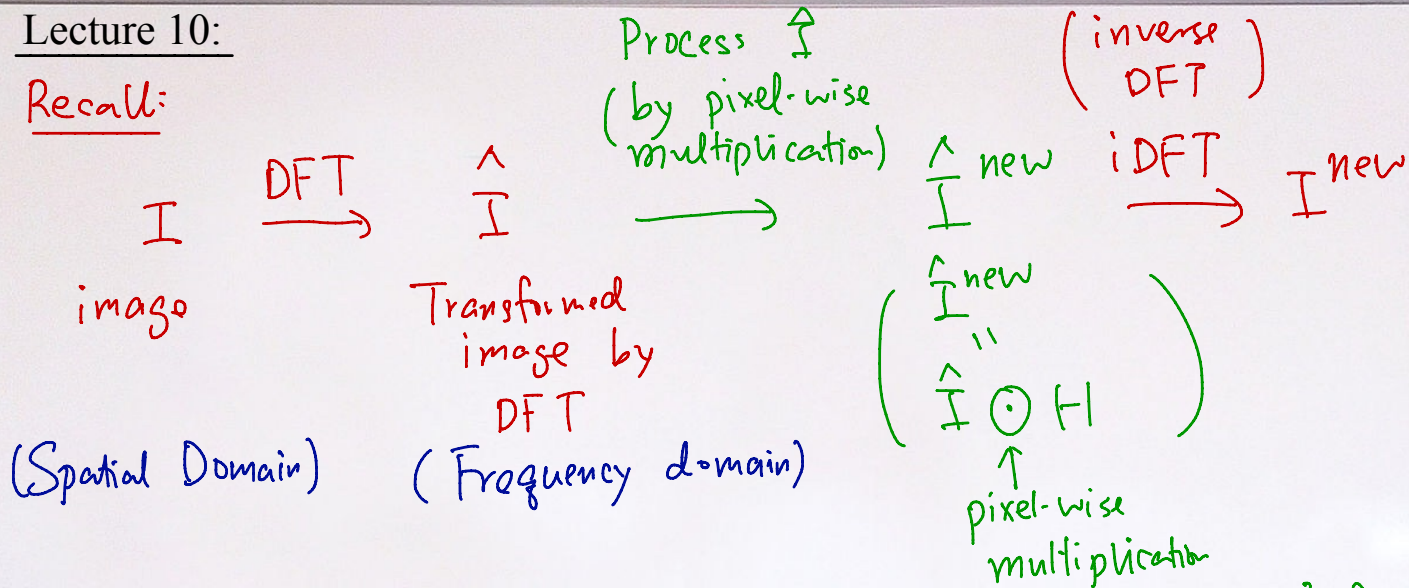


Lecture 10:

Recall:



eg. $\hat{I} \odot H = \begin{pmatrix} 1 & 2 & 3 \\ 4 & 5 & 6 \\ 7 & 8 & 9 \end{pmatrix} \odot \begin{pmatrix} 1 & 1 & 1 \\ 1 & 1 & 0 \\ 0 & 0 & 0 \end{pmatrix} = \begin{pmatrix} 1 & 2 & 3 \\ 4 & 5 & 0 \\ 0 & 0 & 0 \end{pmatrix}$

Understanding convolution:

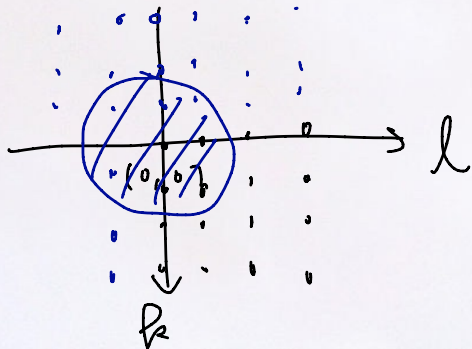
Recall: Discrete convolution:

$$V(n, m) = \sum_{n'=0}^{N-1} \sum_{m'=0}^{N-1} g(n-n', m-m') I(n', m')$$

$g \times I(n, m)$

Linear combination of pixel values of I

In particular, if $g(k, l)$ is only non-zero around $(0, 0)$, then, $g \times I(n, m)$ is a linear combination of pixel value of I around (n, m) !!



Example: Suppose g looks like the following:

$$g = \begin{pmatrix} \text{circle} & 1 & 2 & 3 & \text{circle} \\ 4 & 5 & 6 & & \\ 7 & 8 & 9 & & \end{pmatrix}$$

$\leftarrow R = -1$
 $\leftarrow R = 0$
 $\leftarrow R = 1$

$$I * g(n, m) = \sum_{n'=-1}^{N-1} \sum_{m'=-1}^{M-1} I(n-n', m-m') g(n', m')$$

$\uparrow \quad \uparrow \quad \uparrow$
 $l=-1 \quad l=0 \quad l=1$

Linear combination of neighborhood pixel values

$$\begin{aligned} &= 1 \cdot I(n+1, m+1) + 2 \cdot I(n+1, m) + 3 \cdot I(n+1, m-1) \\ &+ 4 \cdot I(n, m+1) + 5 \cdot I(n, m) + 6 \cdot I(n, m-1) \\ &+ 7 \cdot I(n-1, m+1) + 8 \cdot I(n-1, m) + 9 \cdot I(n-1, m-1) \end{aligned}$$

Note:

(Spatial domain)

$I * g$

(Linear filtering:
Linear combination of
neighborhood pixel
values)

↓ DFT

(Frequency domain)

$MN \hat{I} \odot \hat{g}$
pixel-wise
multiplication

(Modifying the
Fourier coefficients
by multiplication)

Image enhancement in the frequency domain:

- Goal: 1. Remove high-frequency components (low-pass filter) for image denoising.
noise
2. Remove low-frequency components (high-pass filter) for the extraction of image details.
non-edge

High/Low frequency components of \hat{F}

Let F be a $N \times N$ image, $N = \text{even}$. Let $\hat{F} = \text{DFT of } F$.

$$\therefore \hat{F}(k, l) = \frac{1}{N^2} \sum_{m=0}^{N-1} \sum_{n=0}^{N-1} F(m, n) e^{-j2\pi \cdot \frac{(mk+nl)}{N}}$$

↑
Fourier coefficients of F at (k, l)

$$e^{-j\pi m} e^{-j\pi n} e^{-j\frac{2\pi}{N}(mk+nl)}$$

Observe that: for $0 \leq k, l \leq \frac{N}{2} - 1$

$$\begin{aligned} \hat{F}\left(\frac{N}{2} + k, \frac{N}{2} + l\right) &= \frac{1}{N^2} \sum_{m=0}^{N-1} \sum_{n=0}^{N-1} F(m, n) e^{-j\frac{2\pi}{N}\left(m\left(\frac{N}{2} + k\right) + n\left(\frac{N}{2} + l\right)\right)} \\ &= \frac{1}{N^2} \sum_{m=0}^{N-1} \sum_{n=0}^{N-1} F(m, n) (-1)^{m+n} e^{-j\frac{2\pi}{N}(m(-k) + n(-l))} \end{aligned}$$

A hand-drawn equation on a whiteboard. On the left is a complex, irregular waveform representing a signal with noise. This is followed by an equals sign. To the right of the equals sign are three terms added together: a smooth sine wave with a low frequency, a sine wave with a medium frequency, and a sine wave with a high frequency. Each sine wave is preceded by a lowercase letter (a, b, and c) representing its amplitude. The entire equation is written in red ink.

$$\text{Noisy Signal} = a \sin(\omega_1 t) + b \sin(\omega_2 t) + c \sin(\omega_3 t)$$

To remove noise, truncate c (let $c=0$)

$$= \frac{1}{N^2} \sum_{m=0}^{N-1} \sum_{n=0}^{N-1} F(m,n) e^{-j\frac{2\pi}{N}(m(\frac{N}{2}-k) + n(\frac{N}{2}-l))}$$

$$= \hat{F}\left(\frac{N}{2}-k, \frac{N}{2}-l\right)$$

\therefore Computing part of \hat{F} can determine the rest!!

We have:

$$F(m,n) = \sum_{0 \leq k, l \leq \frac{N}{2}-1} \left[\hat{F}\left(\frac{N}{2}+k, \frac{N}{2}+l\right) e^{j\frac{2\pi}{N}[(\frac{N}{2}+k)m + (\frac{N}{2}+l)n]} \right]$$

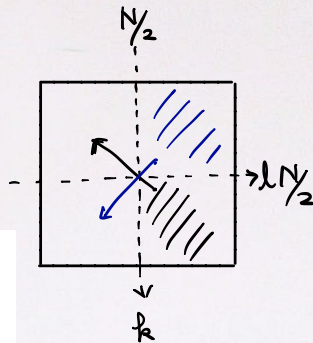
$$+ \sum_{1 \leq k, l \leq \frac{N}{2}-1} \left[\hat{F}\left(\frac{N}{2}+k, \frac{N}{2}+l\right) e^{j\frac{2\pi}{N}[(\frac{N}{2}-k)m + (\frac{N}{2}-l)n]} \right]$$

$$+ \sum_{0 \leq k, l \leq \frac{N}{2}-1} \left[\hat{F}\left(\frac{N}{2}+k, \frac{N}{2}-l\right) e^{j\frac{2\pi}{N}[(\frac{N}{2}+k)m + (\frac{N}{2}-l)n]} \right]$$

$$+ \sum_{1 \leq k, l \leq \frac{N}{2}-1} \left[\hat{F}\left(\frac{N}{2}+k, \frac{N}{2}-l\right) e^{j\frac{2\pi}{N}[(\frac{N}{2}-k)m + (\frac{N}{2}-l)n]} \right]$$

$$+ \sum_{0 \leq l \leq \frac{N}{2}-1} \hat{F}\left(0, \frac{N}{2}+l\right) e^{j\frac{2\pi}{N}[(\frac{N}{2}+l)n]} + \sum_{1 \leq l \leq \frac{N}{2}-1} \overline{\hat{F}\left(0, \frac{N}{2}+l\right)} e^{j\frac{2\pi}{N}[(\frac{N}{2}-l)n]}$$

$$+ \sum_{0 \leq k \leq \frac{N}{2}-1} \hat{F}\left(\frac{N}{2}+k, 0\right) e^{j\frac{2\pi}{N}[(\frac{N}{2}+k)m]} + \sum_{1 \leq k \leq \frac{N}{2}-1} \overline{\hat{F}\left(\frac{N}{2}+k, 0\right)} e^{j\frac{2\pi}{N}[(\frac{N}{2}-k)m]} + \hat{F}(0,0)$$



Observation:

1. When k and l are close to $N/2$, $\hat{F}\left(\underset{\substack{\text{SS} \\ N}}{\frac{N}{2}+k}, \underset{\substack{\text{SS} \\ N}}{\frac{N}{2}+l}\right)$ is associated to $e^{j\frac{2\pi}{N}\left(\left(\frac{N}{2}+k\right)m + \left(\frac{N}{2}+l\right)n\right)}$

\therefore Fourier coefficients at the bottom right are associated to low frequency components!

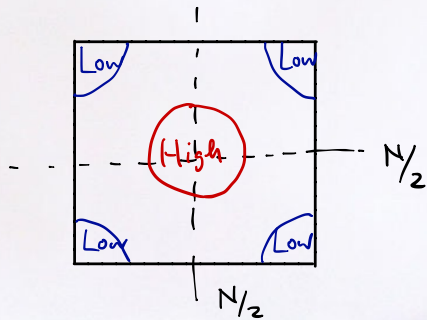
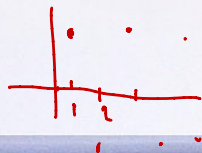
$$e^{j\frac{2\pi}{N}\left(\overset{\text{SS}}{k'm} + \overset{\text{SS}}{l'n}\right)} \quad \text{where } (k', l') \\ \cos\left(\frac{2\pi}{N}(k'm + l'n)\right) + \quad \text{SS} \\ i \sin\left(\frac{2\pi}{N}(k'm + l'n)\right) \quad (0,0)$$

2. Similarly, we can check that Fourier coefficients at the 4 corners are associated to low frequency components.

Low-frequency
if $(k, l) \approx (0,0)$

3. Fourier coefficients in the middle are associated to high-frequency components

$$e^{j\frac{2\pi}{N}\left(\frac{N}{2}m + \frac{N}{2}n\right)} \\ = e^{j\pi(m+n)} = (-1)^{m+n}$$



\therefore High-pass filtering
Remove coefficients at 4 corners
Low-pass filtering
Remove coefficients at the center

Centralisation:

Assume periodic conditions on F .

We can let $\tilde{F}(u, v) = \hat{F}(u - \frac{N}{2}, v - \frac{N}{2})$ where $0 \leq u \leq N-1$
 $0 \leq v \leq N-1$

Then, High-frequency components are located at 4 corners of $\tilde{F}(u, v)$

Low-frequency components are located at center of $\tilde{F}(u, v)$

Mathematically,

Consider the discrete Fourier transform of $(-1)^{x+y}F(x, y)$:

$$\begin{aligned} & \text{DFT}(f(x, y)(-1)^{x+y})(u, v) \quad e^{j2\pi(\frac{N}{2}x + \frac{N}{2}y)} \\ &= \frac{1}{N^2} \sum_{x=0}^{N-1} \sum_{y=0}^{N-1} f(x, y) e^{j\pi(x+y)} \exp\left(-j2\pi\left(\frac{ux}{N} + \frac{vy}{N}\right)\right) \\ &= \frac{1}{N^2} \sum_{x=0}^{N-1} \sum_{y=0}^{N-1} f(x, y) \exp\left(-j2\pi\left(\frac{(u - N/2)x}{N} + \frac{(v - N/2)y}{N}\right)\right) \\ &= \hat{F}\left(u - \frac{N}{2}, v - \frac{N}{2}\right) = \tilde{F}(u, v) \end{aligned}$$

Therefore, to compute $\tilde{F}(u, v)$, we can compute DFT of $(-1)^{x+y}f(x, y)$.

Definition 3.5: A **low-pass filter (LPF)** (LPF) leaves low frequencies unchanged, while attenuating the high frequencies.

A **high-pass filter (HPF)** leaves high frequencies unchanged, while attenuating the low frequencies.

Basic steps of filtering in the frequency domain

1. Multiply $f(x, y)$ by $(-1)^{x+y}$.
2. Compute $\tilde{F}(u, v) = DFT(f(x, y)(-1)^{x+y})(u, v)$.
3. Multiply \tilde{F} by a real "filter" function $H(u, v)$ to get

$$G(u, v) = H(u, v)\tilde{F}(u, v)$$

(point-wise multiplication, but not matrix multiplication)

4. Compute inverse DFT of $G(u, v)$.
5. Take real part of the result in Step 4.
6. Multiply the result in Step 5 by $(-1)^{x+y}$.

Remark: 1. H is taken either to remove high-frequency coefficients/low-frequency coefficients.

$$2. \mathcal{F}^{-1}(G(u,v)) = g(x,y) = \frac{1}{N_x} \mathcal{F}^{-1}(H(u,v)) \times \mathcal{F}^{-1}(\tilde{F}(u,v)) = \frac{1}{N_x} h * \tilde{f}(x,y)$$

Filtering in the frequency domain = Linear filtering in the spatial domain!

Example of Low-pass filters for image denoising

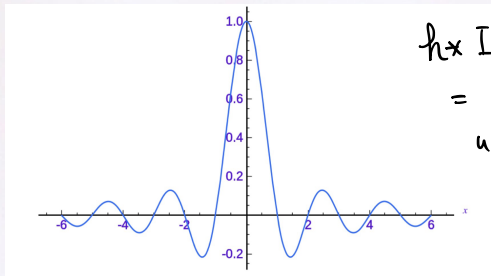
Assume that we work on the centered spectrum!

That is, consider $\hat{F}(u, v)$ where $-\frac{N}{2} \leq u \leq \frac{N}{2} - 1$, $-\frac{N}{2} \leq v \leq \frac{N}{2} - 1$.

1 Ideal low pass filter (ILPF):

$$H(u, v) = \begin{cases} 1 & \text{if } D(u, v) := u^2 + v^2 \leq D_0^2 \\ 0 & \text{if } D(u, v) > D_0^2 \end{cases}$$

In 1-dim cross-section, $\mathcal{F}^{-1}(H(u, v))$ looks like:



$$h \times I(x, y)$$

$$= \sum_{u, v} h(x-u, y-v) I(u, v)$$

every pixel values of I has an effect on $h \times I(x, y)$!!

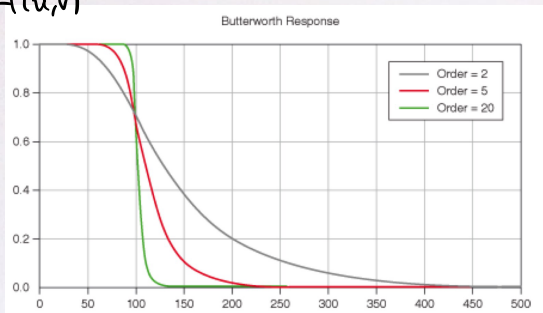
Good: Simple

Bad: Produce ringing effect!

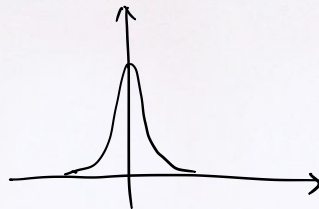
2. Butterworth low-pass filter (BLPF) of order n ($n \geq 1$ integer):

$$H(u, v) = \frac{1}{1 + (D(u, v)/D_0)^n}$$

$H(u, v)$ in 1-dim



$\mathcal{F}^{-1}(H(u, v))$ in 1-dim

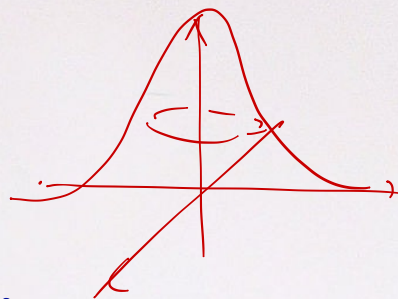


Good: Produce less / no visible ringing effect if n is carefully chosen!!

3. Gaussian low-pass filter

$$H(u, v) = \exp\left(-\frac{D(u, v)}{2\sigma^2}\right)$$

σ = spread of the Gaussian function



F.T. of Gaussian is also Gaussian!!

Good: No visible ringing effect!!

Examples for high-pass filtering for feature extraction

1. Ideal high-pass filter: (IHPF)

$$H(u, v) = \begin{cases} 0 & \text{if } D(u, v) \leq D_0^2 \\ 1 & \text{if } D(u, v) > D_0^2 \end{cases}$$

Bad: Produce ringing

2. Butterworth high-pass filter:

$$H(u, v) = \frac{1}{1 + \left(\frac{D_0}{D(u, v)}\right)^{2n}}$$

($H(u, v) = 0$ if $D(u, v) = 0$)

Choose the right n

Good: Less ringing

3. Gaussian high-pass filter

$$H(u, v) = 1 - e^{-\left(\frac{D(u, v)}{2\sigma^2}\right)^2}$$

Good: No visible ringing!